

SCHOOL OF ELECTRONICS ENGINEERING

Devon Nanaekow Fynn

EE506: Fundamentals of Photonics Assignment

I understand that the University regards breaches of academic integrity and plagiarism as grave and serious. I have read and understood the DCU Academic Integrity and Plagiarism Policy . I accept the penalties that may be imposed should I engage in practice or practices that breach this policy. I have identified and included the source of all facts, ideas, opinions, viewpoints of others in the assignment references. Direct quotations from books, journal articles, internet sources, module text, or any other source whatsoever are acknowledged and the sources cited are identified in the assignment references. I declare that this material, which I now submit for assessment, is entirely my own work and has not been taken from the work of others save and to the extent that such work has been cited and acknowledged within the text of my work. I have read and understood the DCU library referencing guidelines (available at: http://www.library.dcu.ie/LibraryGuides/DCU Library Guide to Harvard Style of Citing & Referencing/player.html Referencing/player.html accessed 9th January 2013) and/or recommended in the assignment guidelines and/or programme documentation. By signing this form or by submitting this material online I confirm that this assignment, or any other course of study. By signing this form or by submitting material for assessment online I confirm that I have read and understood DCU Academic Integrity and Plagiarism Policy (available at: http://www4.dcu.ie/registry/examinations/plagiarism.pdf accessed 9th January 2013)

Programme/Major: Student Number: Date: M.eng Electronic Engineering/Semiconductors 14212148 23rd April 2015

1.1

N = Carrier Number I = biased laser current, A = gain coefficient $N_0 = carrier number at transparency$ P = Photon number $T_e = Carrier lifetime$ $T_p = Photon lifetime$ $\beta = spontaneous emission coupling$

1.2

Conditions for threshold current dN/dt = 0 and dP/dt = 0, $GP >> \beta N/T_e$

The spontaneous emission component $(\beta \frac{N}{T_e})$ is neglected

So:
$$N = N_0 + \frac{1}{AT_p}$$

 $P = \frac{T_p}{e} (I - I_s)$ *I_s* is the threshold current:

$$I_{s} = \frac{e}{T_{e}} \left(N_{0} + \frac{1}{AT_{p}} \right)$$

$$I_{s} = \frac{1.6 \times 10^{-19}}{2.2 \times 10^{-9}} \left(1.2 \times 10^{8} + \frac{1}{8 \times 10^{3} \times 1.6 \times 10^{-12}} \right)$$

$$I_{s} = 0.0144 A$$

1.3

Above threshold: Carrier Number (N) corresponds to the number of carriers when the gain is equal to the losses, P increases with the current but there is no change in N

$$N = N_0 + \frac{1}{AT_p}$$

Below threshold: Gain is zero, so: $A(N - N_0)P = 0$, when this is applied to the carrier number equation

$$N = \frac{T_e}{e}I$$



2.1 The spontaneous emission component $(\beta \frac{N}{T_e})$ is neglected $\frac{dP}{dt} = A(N - N_0)P - \frac{P}{T_p}$ $\frac{dN}{dt} = \frac{I}{e} - A(N - N_0)P - \frac{N}{T_e}$

Applying perturbation condition: $I = \overline{I} + \delta I$, $P = \overline{P} + \delta P$, $N = \overline{N} + \delta N$ So:

$$\frac{d(P+\delta P)}{dt} = A(\overline{N}+\delta N-N_0)(\overline{P}+\delta P) - \frac{P+\delta P}{T_p}$$
$$\frac{d(\overline{N}+\delta N)}{dt} = \frac{\overline{I}+\delta I}{e} - A(\overline{N}+\delta N-N_0)(\overline{P}+\delta P) - \frac{\overline{N}+\delta N}{T_e}$$

So:

$$\frac{d\bar{P}}{dt} + \frac{d\delta P}{dt} = A(\bar{N} - N_0)\bar{P} - \frac{\bar{P}}{T_p} + A(\bar{N} - N_0)\delta P + A\bar{P}\delta N + A\delta N\delta P - \frac{\delta P}{T_p}$$

$$\frac{d\bar{N}}{dt} + \frac{d\delta N}{dt} = \frac{\bar{I}}{e} - A(\bar{N} - N_0)\bar{P} - \frac{\bar{N}}{T_e} + \frac{\delta I}{e} - A(\bar{N} - N_0)\delta P - A\bar{P}\delta N - A\delta N\delta P - \frac{\delta N}{T_e}$$
All 2nd order terms are ignored and $A(\bar{N} - N_0) = \frac{1}{2}$

All 2nd order terms are ignored, and $A(\overline{N} - N_0) = \frac{1}{T_p}$

So:

$$\frac{d\delta P}{dt} = \frac{\bar{P}}{T_p} - \frac{\bar{P}}{T_p} + \frac{\delta P}{T_p} + A\bar{P}\delta N - \frac{\delta P}{T_p}$$
$$\frac{d\delta N}{dt} = \frac{\bar{I}}{e} - \frac{\bar{P}}{T_p} - \frac{\bar{N}}{T_e} + \frac{\delta I}{e} - \frac{\delta P}{T_p} - A\bar{P}\delta N - \frac{\delta N}{T_e}$$

So:

$$\frac{d\delta P}{dt} = A\bar{P}\delta N$$
$$\frac{d\delta N}{dt} = \frac{\delta I}{e} - \frac{\delta P}{T_p} - (A\bar{P} + \frac{1}{T_e})\delta N$$

So:

$$\begin{pmatrix} \frac{d\delta N}{dt} \\ \frac{d\delta P}{dt} \end{pmatrix} = \begin{pmatrix} -(A\overline{P} + \frac{1}{T_e}) & -\frac{1}{T_p} \\ A\overline{P} & 0 \end{pmatrix} \begin{pmatrix} \delta N \\ \delta P \end{pmatrix} + \begin{pmatrix} \delta i/e \\ 0 \end{pmatrix}$$

2.2 Elements of the matrix $M = \begin{pmatrix} -(A\bar{P} + \frac{1}{T_e}) & -\frac{1}{T_p} \\ A\bar{P} & 0 \end{pmatrix}$

$$A = Gain \ coefficient$$
 $\overline{P} = Average \ number \ of \ photon$
 $T_p = Photon \ lifetime$ $T_e = Carrier \ lifetime$

2.3 The signal can be rewritten as $\frac{d}{dt} \begin{pmatrix} \delta N \\ \delta P \end{pmatrix} = \begin{pmatrix} -(A\bar{P} + \frac{1}{T_e}) & -\frac{1}{T_p} \\ A\bar{P} & 0 \end{pmatrix} \begin{pmatrix} \delta N \\ \delta P \end{pmatrix} + \begin{pmatrix} \delta i/e \\ 0 \end{pmatrix}$ Converting to Fourier domain $\begin{pmatrix} j\Omega \delta \widetilde{N} \\ j\Omega \delta \widetilde{P} \end{pmatrix} = \begin{pmatrix} -(A\bar{P} + \frac{1}{T_e}) & -\frac{1}{T_p} \\ A\bar{P} & 0 \end{pmatrix} \begin{pmatrix} \delta \widetilde{N} \\ \delta \widetilde{P} \end{pmatrix} + \begin{pmatrix} \delta i/e \\ 0 \end{pmatrix}$ $\begin{pmatrix} j\Omega + (A\bar{P} + \frac{1}{T_e}) & \frac{1}{T_p} \\ -A\bar{P} & j\Omega \end{pmatrix} \begin{pmatrix} \delta \widetilde{N} \\ \delta \widetilde{P} \end{pmatrix} = \begin{pmatrix} \delta i/e \\ 0 \end{pmatrix}$

Assuming, $M = \begin{pmatrix} j\Omega + (A\bar{P} + \frac{1}{\tau_e}) & \frac{1}{\tau_p} \\ -A\bar{P} & j\Omega \end{pmatrix}$, $V = \begin{pmatrix} \delta \tilde{N} \\ \delta \tilde{P} \end{pmatrix}$ and $B = \begin{pmatrix} \delta \tilde{\iota/e} \\ 0 \end{pmatrix}$ $V = M^{-1}B$

$$M^{-1} = \frac{1}{\left(\left(j\Omega + \left(A\overline{P} + \frac{1}{T_e}\right)\right) * J\Omega\right) - \left(\frac{1}{T_p} * -A\overline{P}\right)} \begin{pmatrix} j\Omega & -\frac{1}{T_p} \\ A\overline{P} & j\Omega + \left(A\overline{P} + \frac{1}{T_e}\right) \end{pmatrix}$$

N:B: $\frac{AP}{T_p} = \omega_r^2$, $(AP + \frac{1}{T_e}) = \frac{2}{T_r}$, $M^{-1} = \frac{1}{-\Omega^2 + J\Omega \frac{2}{T_r} - \omega_r^2} \begin{pmatrix} j\Omega & -\frac{1}{T_p} \\ \omega_r^2 T_p & j\Omega + \frac{2}{T_r} \end{pmatrix}$ Since: $\begin{pmatrix} \delta N \\ \delta P \end{pmatrix} = M^{-1} \begin{pmatrix} \delta i/e \\ 0 \end{pmatrix}$

Then:

$$\widetilde{\delta N} = \frac{j\Omega}{-\Omega^2 + J\Omega \frac{2}{T_r} - \omega_r^2} \frac{\widetilde{\delta \iota}}{e} \qquad and \qquad \widetilde{\delta P} = \frac{\omega_r^2 T_p}{-\Omega^2 + J\Omega \frac{2}{T_r} - \omega_r^2} \frac{\widetilde{\delta \iota}}{e}$$

3.1

$$v = \frac{c}{\gamma} = \frac{3 \times 10^8}{1550 \times 10^{-9}} = 1.94 \times 10^{14}, \ P_s = 501 \times 10^{-6} W$$

$$\frac{(\frac{0.85 \times 1.6 \times 10^{-19}}{6.62 \times 10^{-34} \times 1.94 \times 10^{14}})^2 * (501 \times 10^{-6})^2}{\left(\left((2 \times 1.6 \times 10^{-19}) * \left(\frac{0.85 \times 1.6 \times 10^{-19}}{6.62 \times 10^{-34} \times 1.94 \times 10^{14}}\right) * 501 \times 10^{-6}\right) + \frac{4 \times 1.38 \times 10^{-23} \times 300}{50}\right) * 20 \times 10^9}$$

$$\frac{S}{N} = 44.5 \ dBm$$

3.2 If temperature is at $0^{\circ}c = 273k$

$$\frac{S_{N}}{N} = 45 \ dBm$$

$$\frac{(\frac{0.85 \times 1.6 \times 10^{-19}}{1.3 \times 10^{-19}})^{2} * (1.737 \times 10^{-4})^{2}}{((2 \times 1.6 \times 10^{-19}) * (\frac{0.85 \times 1.6 \times 10^{-19}}{1.3 \times 10^{-19}}) * 1.737 \times 10^{-4}) + \frac{4 \times 1.38 \times 10^{-23} \times 273}{50}) * 20 \times 10^{9}}$$

3.3

Temperature increase would result in decrease in the value of the signal-noiseratio.

The temperature increases produces more noise which results in a lower signalnoise-ratio.

3.4

Bandwidth increase would result in decrease in the value of the signal-noiseratio.

As the noise spreads out over all frequencies, the wider the bandwidth of the device, the greater the noise level. And an increase in noise level results in a lower signal-noise-ratio.

Internal Quantum Efficiency: n_{int}

Carrier lifetime:
$$\tau = \frac{\tau_r \tau_{nr}}{\tau_r + \tau_{nr}} = \frac{50 \times 75}{50 + 75} = 30ns$$
$$n_{int} = \frac{\tau}{\tau_r} = \frac{30}{50} = 0.6$$

Extraction efficiency: n_e

$$n_e = \left[\frac{1}{(n+1)^2}\right] \times \frac{1}{n}$$
$$n_e = \left[\frac{1}{(3.6+1)^2}\right] \times \frac{1}{3.6}$$
$$n_e = 0.013$$

External Quantum Efficiency: n_{ext}

$$n_{ext} = n_e \times n_{int}$$
$$n_{ext} = 0.013 \times 0.6$$
$$n_{ext} = 0.0079$$